

### 1. Absolute Maximum Ratings(Ta=25°C)

Symbol	Terms	Values		Units
		222W	222WA	
V <sub>S</sub>	Supply voltage primary	+16		V
V <sub>in</sub>	Input PWM level(HIGH) V <sub>in</sub> A; V <sub>in</sub> B	V <sub>S</sub> +0.3		V
V <sub>iH</sub>	Input signal voltage(HIGH) (mode select; external error signal)	V <sub>S</sub> +0.3		V
I <sub>oc</sub>	Output signal voltage(HIGH) (error signal output current)	±10		mA
I <sub>outPEAK</sub>	Output peak current	±10	±15	A
I <sub>outAV</sub>	Output average current	±50		mA
V <sub>CES</sub>	Collector-emitter voltage sense	1700		V
V <sub>isol IO</sub>	Isolation test voltage IN-OUT(1 min,ac)	4000		V <sub>ac</sub>
R <sub>G min</sub>	Minimal gate resistance	3	2	Ω
Q <sub>out/pulse</sub>	Charge per pulse	±10	≥10 <sup>(1)</sup>	μC
dv/dt	Rate of rise and fall of voltage	50		kV/μs
f <sub>SW max.</sub>	Maximum operating frequency	50		kHz
I <sub>DC max.</sub>	Maximum DC/DC current primary	280		mA
t <sub>TD min</sub>	Top-bottom interlock dead-time adjusted from factory	10		μs
T <sub>op</sub>	Operating temperature	-40 ... +85		°C
T <sub>stg.</sub>	Storage temperature	-40 ... +85		°C



### POWER-SEM Thick-film Dual IGBT Driver PSHI 222W PSHI 222WA

#### Features

- Thick-film dual IGBT driver
- Drive all series IGBTs with V<sub>CE</sub> up to 1200V & 1700V
- Half-bridge or two independent single drives mode selected
- Short circuit protection by V<sub>CEsat</sub> monitoring
- Soft short circuit turn-off
- Isolation due to ferrite transformer
- Supply undervoltage protection <13.5V
- Error memory
- Driver interlock top/bottom in half-bridge mode
- Internal isolated DC/DC power supply
- High peak output current
- ±15V IGBT gate driving voltage
- Short signal transition time
- Short-pulse control function (<500ns restrained)

#### Typical Applications

- Single and bridge circuit
- Inverter
- Welding machine
- Induction heating
- Converter
- High power UPS
- High frequency SMPS

### 2. Electrical Characteristics(Ta=25°C)

Symbol	Terms	Min.	Typ.	Max.	Rec.	Units
V <sub>S</sub>	Supply voltage primary	+14	+15	+16	+15	V
I <sub>DC</sub>	Supply current primary	100	---	300		mA
V <sub>IT</sub>	Input threshold voltage		8			V
R <sub>in</sub>	Input resistance		10			kΩ
V <sub>G(on)</sub>	Turn-on output gate voltage		+15			V
V <sub>G(off)</sub>	Turn-off output gate voltage		-15			V
t <sub>d(on)IO</sub>	Input-output turn-on propagation time		1			μs
t <sub>d(off)IO</sub>	Input-output turn-off propagation time		1			μs
t <sub>d(Err)</sub>	Error input-output propagation time		650			ns
t <sub>md</sub>	Narrow pulse restrain		500			ns
V <sub>CEstat</sub>	Reference voltage for V <sub>CE</sub> monitoring	2	5.2	6.3	6.3 <sup>(2)</sup>	V
t <sub>BK</sub>	Both low reset when: input signal V <sub>in</sub> A / V <sub>in</sub> B both low		1			ms
C <sub>PS</sub>	Primary to secondary capacitance		16			pF

1, PSHI 222WA can increase this value by adjusting Pins group 19,21,37 & 29, 31,38, which depending on the capacitance of external capacitors.

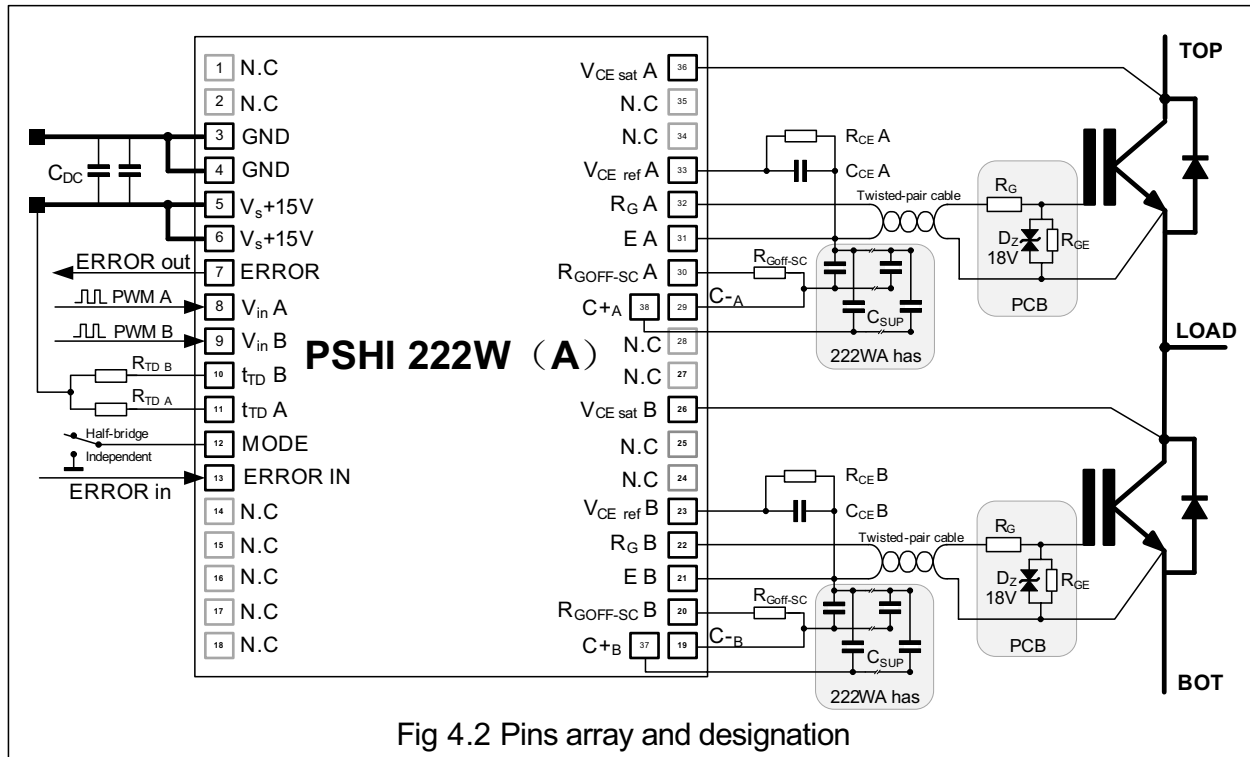
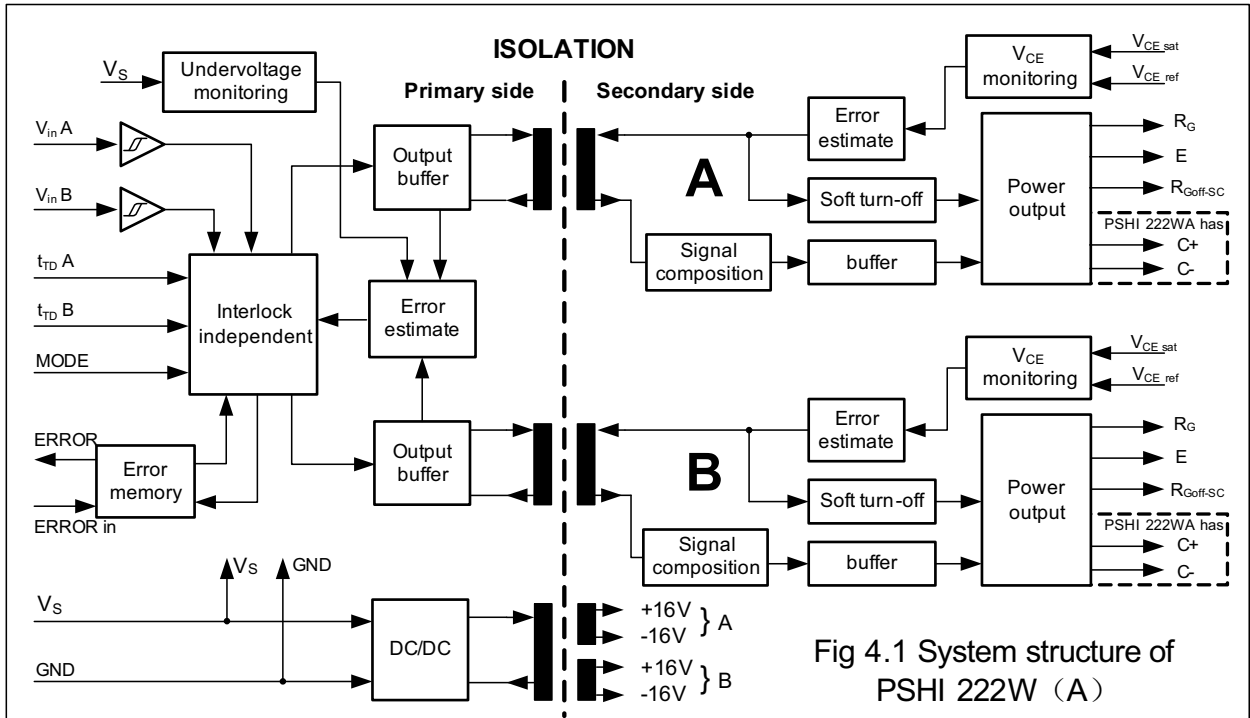
2, 6.3V suit for 1700V IGBT, adjusted from factory; 5.2V suit for 600V & 1200V IGBT.

Max. switching frequency:

$$f_{SW \max.} = \frac{I_{outAV}(\text{mA})}{Q_G(\mu\text{C})}$$

f<sub>sw max.</sub>: maximum switching frequency  
 Q<sub>G</sub>: maximum IGBT gate charge at ΔV<sub>GE</sub>  
 I<sub>outAV</sub>: average cont. output current per channel

System structure of PSHI 222W /PSHI 222WA



### 3. Product Overview

PSHI 222W is a standard IGBT driver for all power IGBTs in the market which drives all IGBTs with  $V_{CE}$  up to 1200V and 1700V. To protect the driver against moisture, dust and salt fog, it adopts module filling package. The adaption of the drivers to the application has been improved by using pins to change several parameters and functions. It can drive different IGBTs by adjusting their gate resistance  $R_G$  and "soft turn-off" resistor  $R_{GOFF-SC}$  simply.

It can drive two IGBTs under half-bridge mode or independent mode. In order to improve driver's anti-jamming ability, it chooses 15V level as control signal, as well as whole logic process. Integrated transformer is divided into three parts: 2 pulse transformers and dual DC/DC switch mode power supply. All of these design can make it supply minimum coupling capacitance and high insulated voltage.

Short turn-off function is provided to protect IGBT from damage by high voltage spike. In case of short circuit, it automatically increases the IGBT turn-off time and hence reduces the  $V_{CE}$  voltage overshoot to improve the IGBT's reliability which make it can be used in higher DC voltage application.

Integrated DC/DC power supply with high galvanic isolation (4 kV/1 minute) ensures that the user is protected from the high voltage (secondary side). The power supply for the driver may be the same as used in the control board (+15V) without the requirement of isolation. All information that is transmitted between input and output uses ferrite transformers, resulting in high dv/dt immunity (50 kV/ $\mu$ s) which has high anti-jamming ability.

### 4. Block diagram PSHI 222W

System structure and performance features: (see fig. 4.1)

#### 4.1 Signal level

The maximum value of PSHI 222W input level is  $V_S+0.3V$ , the reverse short peak current with the same value will not damage input stage. The driver's signal input resistance is 10k $\Omega$ , internal Schmitt trigger upset threshold on input stage is 8V (to GND). Active-logic control whatever under independent mode or half-bridge mode, that is: high level IGBT switch on, while low level IGBT switch off.

Internal narrow pulse restrain circuit, make sure that the narrow pulse <500ns will be restrained.

When connecting PSHI 222W to control board using short connecting lead, otherwise, shielded cable or external interference suppression nets is needed.

#### 4.2 Logic level and error reset

Logic level:

"Logic level" includes: error input/output, mode select and error reset.

The max. value of Input signal level (error output &

mode select) is  $V_S+0.3V$ , upset threshold is 8V. So 15V is very suitable as logic signal.

Mode select see: fig. 4.2

Error reset: After faults detecting, the driver can be activated to reset by adjusting 2 PWM signals-- $V_{in A}$  /  $V_{in B}$ . If install both of PWM signals LOW at the same time for more than 1ms, error reset automatically. See fig. 4.3

Logic output:

PSHI 222W can monitor short-circuit, over-current and undervoltage of IGBT by direct measurement of  $V_{CEsat}$ . It blocks IGBT's driving signals and "soft turn-off" the IGBT in case of faults occurs and sends a error signal to primary side. "Error memory" circuit blocks all pulse output and keeps reset only be possible when both PWM signals are low level for more than 1ms. The low level error signal is fed to terminal by Pin 7. The error output of Pin 7 is open-collector output with internal 10k $\Omega$  pull-up resistor, and the maximum input current is 20mA. The IGBT drivers can use common error output line, no need to connect with pull-up resistor. (See fig. 4.3)

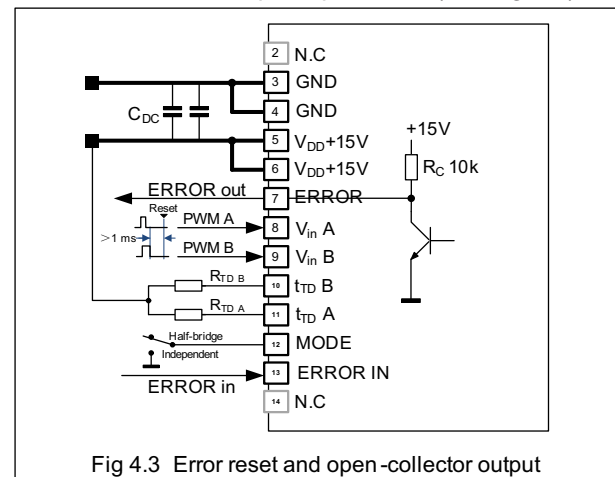


Fig 4.3 Error reset and open-collector output

Additionally, PSHI 222W has another function of external error input. It blocks all IGBT's driving signals output when Pin 13 receiving external input low level error signals. The external error input will not activate "error memory", and blocks will be relieved after it disappearing. Pin 7 & pin 13 of more drivers can be connected together to realize synchronous failure.

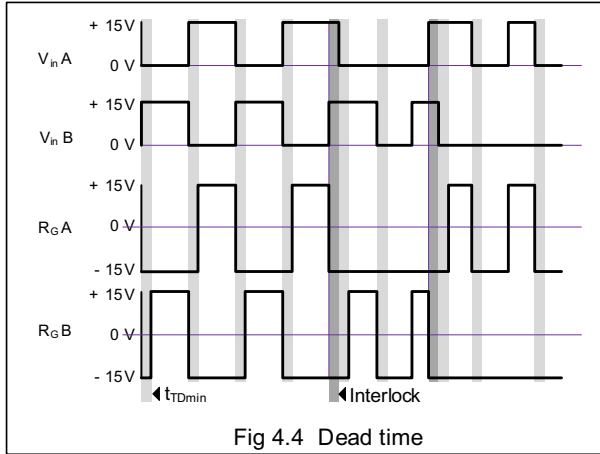
#### 4.3 Working mode select

PSHI 222W has two methods to drive IGBT, including: direct independent mode & half-bridge interlock mode.

Direct independent mode: Under such mode, there is no electrical connection between two channels of PSHI 222W. Channel A & B can work independently, and can be switched on simultaneously. Direct mode can be activated by connecting Pin 12 "mode select" to GND (eg. connect to Pin 3/4). Then  $t_{TD A}$  (Pin 11) &  $t_{TD B}$  (Pin 10) not connected.

Half-bridge interlock mode: It can interlock two channels of PSHI 222W to prevent the top & bottom

IGBTs from being in the on-state simultaneously, which cause an interlock time. The dead time can be set by adjusting external resistor of  $t_{TD A}$  (Pin 11) and  $t_{TD B}$  (Pin 10). Half-bridge interlock mode can be activated by Pin 12 "mode select" not bridged or connecting with  $V_s$  (Pin 5/6). Dead time to see fig. 4.4



Interlock dead time; Under half-bridge mode, there is a minimum inner interlock dead time between 2 channels of PSHI 222W, which adjusted from factory is 10µs. It can be reduced by paralling external resistor to  $V_s$  at  $t_{TD A}$  (Pin 11) &  $t_{TD B}$  (Pin 10) and inner capacitor. Table 4.5 shows the corresponding relationship between  $R_{TD}$  and dead time:

$R_{TD}$	Dead time $t_{TD}$	$R_{TD}$	Dead time $t_{TD}$
10kΩ	0.9µs	68kΩ	4µs
22kΩ	1.8µs	100kΩ	5µs
33kΩ	2.5µs	330kΩ	7.7µs
47kΩ	3.2µs	not connected	10µs

Tab 4.5 Corresponding relationship between  $t_{TD}$  and  $R_{TD}$

The dead time error is decided by external capacitance, so it's better to choose resistors with less error.

**4.5 Power supply**

With internal DC/DC isolation power supply, PSHI 222W can create ±16V voltage on secondary side. Therefore, PSHI 222W only needs a +15V external power supply. In order to work under stable +15V (±1V) power voltage, an additional constant voltage capacitor  $C_{DC}$  (see fig.4.2) with min. capacitance of 20µF is required.

PSHI 222W has a function to monitor under-voltage of primary side. If power voltage of primary side below +13.5V, an error signal can be created to turn-off the driver and sends a low level error signal to outside by Pin 7.

**4.6 Connecting IGBT**

PSHI 222W can drive two single IGBTs by two independent channels, also can drive an IGBT's half-bridge by half-bridge mode. The size of PSHI 222WA (auxiliary board) is depending on IGBT's input capacitance and switching frequency.

When considering gate current, please note that it's

not only determined by external gate resistance, but also need to take the internal gate resistance of IGBT into consideration. That is because most of high power IGBTs already have an integrated gate resistor inside (refer to relevant IGBT datasheets). Additionally, PSHI 222W has low output resistance. So it's not practical to calculate gate current only by external gate resistor. The approximate value of driving power and peak current can be calculated with:

$$P_G = f_{sw} \times \Delta V_{GE}^2 \times C_{ies} \times 4$$

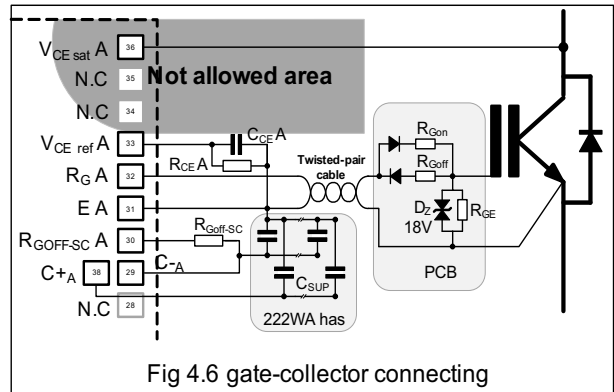
$$P = P_G + P_S$$

$$\text{Max. driver current } I_{G \text{ MAX}} = \frac{\Delta V_{GE}}{I_{G \text{ min}}}$$

- $f_{sw}$  = switching frequency
- $C_{ies}$  = input capacity (datasheet)
- $P_S$  = driver dissipation
- $\Delta V = 30V @ \pm 15V$
- $R_{G \text{ min}} = R_{G \text{ extern}} + R_{G \text{ intern}}$

**Connecting gate:**

IGBT gate is connected with  $R_G A$  or  $R_G B$  of the driver by external gate resistor  $R_G$ . Connect the auxiliary emitter of IGBT with E A or E B output of PSHI 222W directly. The gate output voltage of the driver  $V_G$  is ±15V, and the selection for external gate resistor  $R_G$  is according to the relevant IGBT (refer to the datasheets). Apart from gate resistor  $R_G$ , a gate-emitter resistor  $R_{GE}$  and gate clamp are also needed. Better to keep  $R_{GE} \leq 10k\Omega$ , and gate clamping can be realized by zener diode or TVS diode. Both the diodes should keep breakthrough voltage  $\leq 18V$ , which can protect gate voltage from rising up severely because of parastatics effect (eg. Miller Effect). When connecting PSHI 222W to IGBT gate, flat cable with pairs of conductors twisted is highly recommended and should be kept as short as possible (usually shorter than 20cm). Also, please fix gate resistor  $R_G$ , voltage clamp diode  $D_z$  and gate-emitter resistor  $R_{GE}$  onto a small PCB. The gate PCB must be installed closed to IGBT.



As a general rule, we only need 1 gate resistor  $R_G$  to drive IGBT. If control di/dt or dv/dt of IGBT turn-on and turn-off by rising and dropping gate driving signals, please separate gate driving signals. See fig.4.6.

Connecting collector:

$V_{CE\ sat}$  monitoring,

" $V_{CE\ sat}$  monitoring circuit" is responsible for short-circuit sensing. Due to the direct measurement of  $V_{CE\ stat}$  on the IGBT's collector, it blocks the output buffer (through the soft turn-off circuit) in case of short-circuit and sends a signal to the ERROR memory on the primary side.

Attention: Due to the connection to high voltage with  $V_{CE\ sat}$  monitor terminal of Pin 26&36, there must be no any other components or wirings over the 6mm radius area enclosed by  $V_{CE\ sat}$  monitor pins on auxiliary board and connecting leads in order to avoid creepage. See fig. 4.6.

4.7 IGBT short-circuit and "soft turn-off"

In case of short-circuit, a "soft turn-off" circuit turns-off the IGBT at a lower speed by increasing the turn-off resistance. The IGBT will be turned-off after time  $t_{SC}$ . This produces a smaller voltage spike above the DC link by reducing di/dt value. Because in short-circuit conditions, the Homogeneous IGBT's peak current increases up to 6-8 times the IGBT's rated current, and some stray inductance is always present in power loop circuits, it must fall to zero in a longer time than at normal operation to avoid damage IGBT by high voltage spike.

The internal resistor used for soft turn-off is 22Ω, this "soft turn-off" time can be reduced by paralleling resistor  $R_{goff-SC}$  on Pin 19,20 & 29,30, with internal resistor of PSHI 222W.

4.8 Reference voltage  $V_{CE\ ref}$

The reference voltage  $V_{CE\ ref}$  adjusted dynamically according to IGBT switch characteristics, and reset when IGBT turn-off. The  $V_{CE\ ref}$  is not static but a dynamic reference which has an exponential shape starting at about 15 V and decreases to  $V_{CE\ stat}$  (determined by  $R_{CE}$ ), with a time constant  $T$  (controlled by  $C_{CE}$ ) (see Fig.4.7).

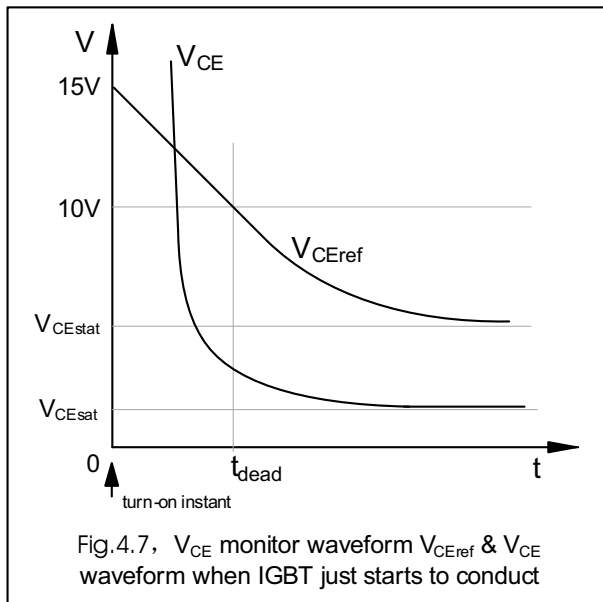


Fig.4.7,  $V_{CE}$  monitor waveform  $V_{CE\ ref}$  &  $V_{CE}$  waveform when IGBT just starts to conduct

$V_{CE\ stat}$  threshold is a static value of  $V_{CE\ ref}$  which is controlled by resistor  $R_{CE}$ . It can be adjusted by resistor  $R_{CE}$  to reach the maximum value as per IGBT's demand.  $V_{CE\ stat} > V_{CE\ sat}$  under normal conditions, but will not exceed 10V. The decay time of  $V_{CE\ ref}$  is determined by capacitor  $C_{CE}$  and resistor  $R_{CE}$  (see Fig.4.7). It controls the dead time  $t_{dead}$  when IGBT just starts to conduct till  $V_{CE\ sat}$  monitoring starts. To avoid a false failure indication when the IGBT just starts to conduct ( $V_{CE} > V_{CE\ ref}$ ), some decay time  $t_{dead}$  must be provided for the  $V_{CE\ ref}$ . As the  $V_{CE}$  signal is internally limited at 10V, " $V_{CE}$  monitoring circuit" will be trigger and cut off IGBT by "soft turn-off circuit". when  $V_{CE\ ref}$  drops to 10V (ie. leave monitoring dead area  $t_{dead}$ ) and  $V_{CE}$  voltage rises above the reference voltage at any time ( $V_{CE} > V_{CE\ ref}$ ). The various different operating conditions are depicted in fig.4.8.

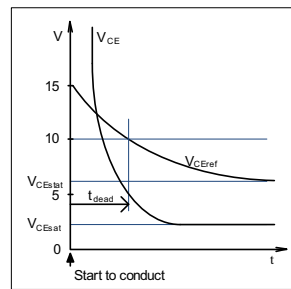


Fig 4.8a, Usual case

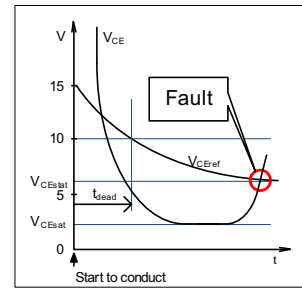


Fig 4.8b, Short circuit during operation

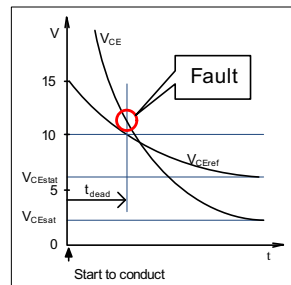


Fig 4.8c, IGBT turns on too slowly or dead time is too short

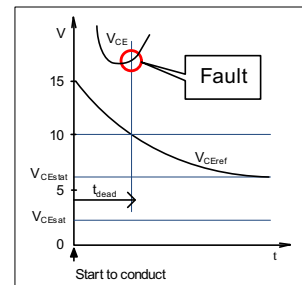


Fig 4.8d, Short circuit during turn-on

The monitor sensitivity of " $V_{CE}$  monitoring circuit" is adjusted by changing dead time  $t_{dead}$ . This can be realized by extending or reducing monitoring dead area via adjusting  $C_{CE}$ . Please make sure the total time from IGBT conduct (start from short circuit) to IGBT totally turn-off by soft turn-off circuit must shorter than IGBT safe soft circuit time (usually 10us or 6us, details pls refer to IGBT supplier). The total time includes  $t_{dead}$ ,  $t_{d( err )}$ ,  $t_{SC}$ , IGBT turn-off trailing time and safety time.

The internal  $V_{CE\ ref}$  of PSHI 222W (adjusted from factory) is  $R_{CE} = 36k\Omega$ ,  $C_{CE} = 330pF$ , then  $V_{CE\ stat} = 6.3V$ ,  $t_{dead} = 3.3\mu s$ . In real application, suitable value can be adjusted by paralleling external  $R_{CE}$ ,  $C_{CE}$  between  $V_{CE\ ref}$  (Pin 23,33) & E (Pin 21,31) with internal resistors and capacitors. See fig.4.6.

1200V IGBT suggest to be connected with a 36kΩ resistor, then  $V_{CEstat} = 5.2V$ ;  $t_{dead} = 3.3\mu s$ .  
 1700V IGBT suggest to be connected with a 140pF capacitor, then  $V_{CEstat} = 6.3V$ ;  $t_{dead} = 4.4\mu s$ .

**4.9 Additional isolation power supply output and snubber capacitor (PSHI 222WA)**

In order to avoid voltage fluctuation by high current,PSHI 222WA has +16V power supply C+pins access. User can also get ±16V isolation power supply between C+/C- and E by using internal DC/DC isolation power supply on secondary side. Above-mentioned E is simulated as GND to IGBT emitter. To avoid voltage sag because of high current,PSHI 222WA must increase an external snubber capacitor  $C_{SUP}$  between C+/C- and E in real application. Snubber capacitor  $C_{SUP}$  should be closed to driver's pins. Due to DC/DC working at frequency of 500kHz,high-frequency low-loss MLCC(multilayer ceramic capacitor) with high ripple current must be used as snubber capacitor. The capacitance is depending on output current, each channel needs

20μF at least in general. Snubber capacitor  $C_{SUP}$  is recommended to be fixed onto the under parts of driver's inner side in order to avoid reducing isolation grade for routing reason. See fig.4.9

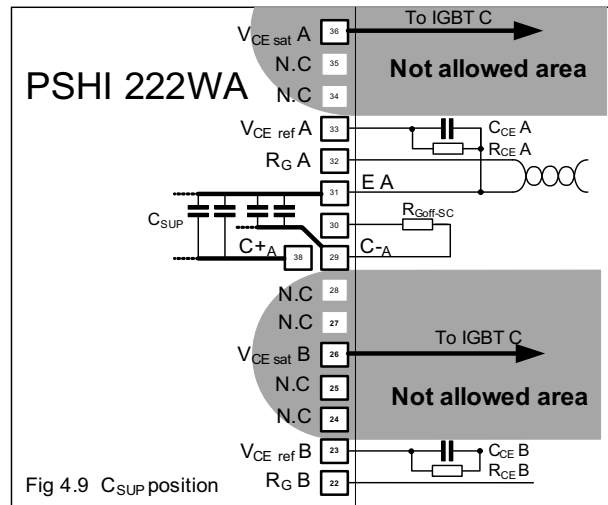


Fig 4.9 C<sub>SUP</sub> position

**5. Pins function and description**

**5.1 Pins array and designation refer to fig.4.2**

**5.2 PSHI 222W, PSHI 222WA pins function**

PIN NO.	Designation	Function	PIN NO.	Designation	Function
1	N.C	Physically non existent	19	C- B	BOT -16V isolation power supply, connect with cathode for snubber capacitor
2	N.C	Physically non existent	20	R <sub>Goff-SC</sub> B	BOT "soft turn-off" resistor
3	GND	ground	21	E B	BOT supply-ground & signal-ground, connect with IGBT auxiliary emitter
4	GND	ground	22	R <sub>G</sub> B	BOT gate driving signal output, connect with IGBT gate by external gate resistor
5	V <sub>S</sub> +15V	supply voltage, +15V	23	V <sub>CE ref</sub> B	BOT R; C net, set V <sub>CEref</sub> curve
6	V <sub>S</sub> +15V	supply voltage, +15V	24	N.C	Physically non existent
7	ERROR	error signal, open-collector output, low level active	25	N.C	Physically non existent
8	V <sub>in</sub> A	TOP PWM input + 15V level	26	V <sub>CE sat</sub> B	BOT V <sub>CE sat</sub> monitoring input, connect with IGBT collector
9	V <sub>in</sub> B	BOT PWM input + 15V level	27	N.C	Physically non existent
10	t <sub>TD</sub> B	BOT interlock dead time setting (half-bridge mode active)	28	N.C	Physically non existent
11	t <sub>TD</sub> A	TOP interlock dead time setting (half-bridge mode active)	29	C- A	TOP -16V isolation power supply, connect with cathod for snubber capacitor
12	MODE	working mode select, GND: independen; Not bridged: half-bridge	30	R <sub>Goff-SC</sub> A	TOP "soft turn-off" resistor
13	ERROR in	External error input, low level active	31	E A	TOP supply-ground & signal-ground, connect with IGBT auxiliary emitter
14	N.C	Physically non existent	32	R <sub>G</sub> A	TOP gate driving signal output, connect with IGBT gate by external gate resistor
15	N.C	Physically non existent	33	V <sub>CE ref</sub> A	TOP R; C net, set V <sub>CEref</sub> curve
16	N.C	Physically non existent	34	N.C	Physically non existent
17	N.C	Physically non existent	35	N.C	Physically non existent
18	N.C	Physically non existent	36	V <sub>CE sat</sub> A	TOP V <sub>CE sat</sub> monitoring input,connect with IGBT collector
			37	C+ B (22WA)	BOT +16V isolation power, connect with anode for snubber capacitor
			38	C+ A (22WA)	TOP +16V isolation power, connect with anode for snubber capacitor

### 6. Mounting dimensions

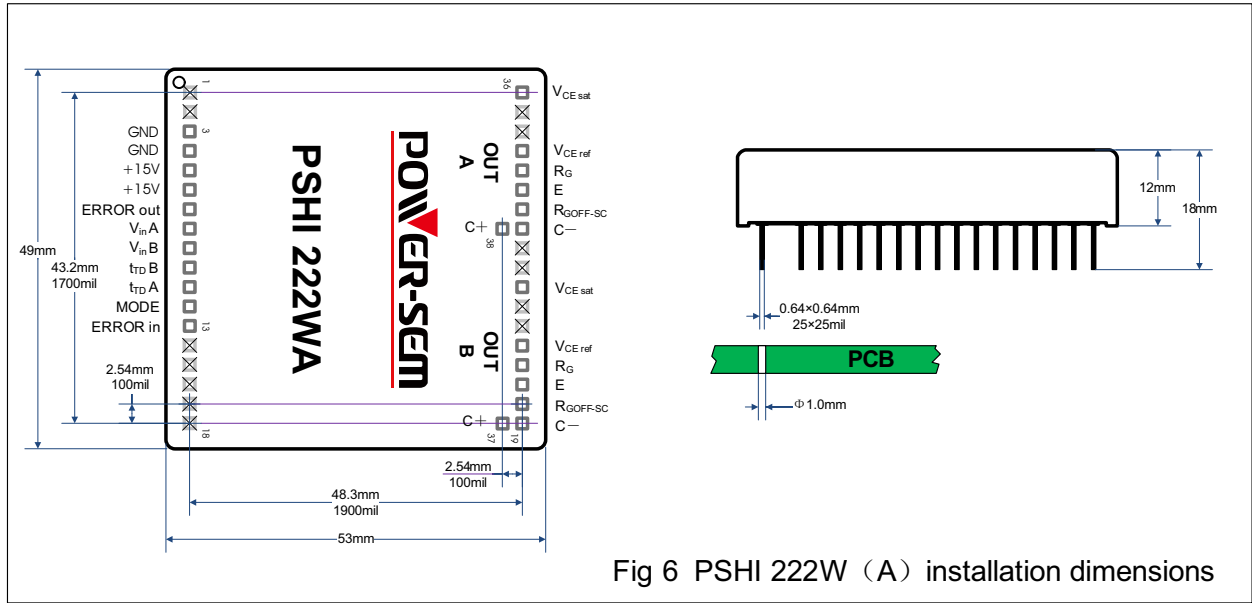


Fig 6 PSHI 222W (A) installation dimensions